PERFORMANCE ANALYSIS OF THE CEPSTRAL TECHNIQUE FOR FREQUENCY-DOMAIN OPTICAL-COHERENCE TOMOGRAPHY

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ABSTRACT
Recently, we proposed a noniterative cepstral technique for exact signal recovery in frequency-domain optical-coherence tomography. In this paper, we address the influence of measurement noise on the performance of the method. We derive analytical expressions for the bias and variance of the tomogram under a small noise approximation, and show that our technique yields unbiased and consistent estimators, which have a variance that is proportional to that of the noise and inversely proportional to the data size. We present simulation results to confirm the theoretical derivations. We also derive approximate Cramér-Rao bounds (CRBs) on the achievable accuracy of reconstruction.

Index Terms— frequency-domain optical-coherence tomography (FDOCT), cepstrum, bias, variance, mean square error, Cramér-Rao bound.

1. INTRODUCTION
Frequency-domain optical-coherence tomography (FDOCT) is now widely recognized as a fast technique for non-invasive three-dimensional imaging of biological specimens. With a penetration depth of 2-3 mm in tissue, resolution of the order of a few micrometers, higher sensitivity [1] and faster imaging speed [2], it has superseded its time-domain counterpart. The quality of tomograms that one achieves is closer to that obtained by high-resolution and invasive methodologies such as histology. FDOCT has been successfully applied in tissue-imaging, dermatology, and ophthalmology [2, 3, 4, 5], with some of them being in vivo.

In FDOCT, the depth information is acquired within a single exposure by means of spectrometer detection. A standard approach to reconstruct the tomogram is to apply the inverse Fourier transform [6]. However, it is known to introduce autocorrelation artifacts and complex-conjugate ambiguity. The latter can be suppressed by placing the zero-delay plane of the interferometer outside the specimen. In our earlier work [7], we proposed a new algorithm to suppress the remaining autocorrelation artifacts. In this paper, we analyze its noise sensitivity, derive expressions for the bias and variance and also corroborate the analytical predictions with Monte-Carlo analysis.

2. SIGNAL ACQUISITION AND ARTIFACT-FREE TOMOGRAM RECONSTRUCTION
Figure 1 displays a typical FDOCT experimental setup. It is a Michelson interferometric configuration comprising an object arm and a reference arm. The light source is chosen to possess a broad spectrum so as to achieve high axial resolution. The source output is split into two beams, one of which is directed towards the object arm and the other towards the reference arm. A broadband mirror placed in the reference arm acts as a perfect reflector. The backscattered light from the specimen placed in the object arm is coherently amplified by...
the reference-arm signal, when the two are suitably combined. The resulting interference signal is recorded by a spectrometer, as a function of the wavelength \( \lambda \). These measurements are then mapped onto the wavenumber scale \( k = \frac{2\pi}{\lambda} \). The measured signal then takes the form

\[
I(\omega) = S(\omega) |1 + A(\omega)|^2, \tag{1}
\]

where \( A(\omega) = \int_0^\infty a(z)e^{-i\omega z}dz \), \( \omega = -2kn \), \( n \) being the refractive index of the specimen, and where \( a(z) \) is the scattering function. The objective is to recover \( a(z) \) from \( I(\omega) \). In [7], we proposed a new technique to accomplish this task. It is summarized as follows: Given the measurements \( I(\omega) \) and \( S(\omega) \), we compute the cepstrum \( c(z) = \mathcal{F}^{-1}\{\log(I/S)\}(z) \), where \( \mathcal{F} \) denotes the Fourier transform operator. The causal part of \( c(z) \) is retained by multiplying with a unit-step function \( u(z) \) to yield \( c^+(z) \). Next, we compute the Fourier transform of \( c^+(z) \) to obtain \( C^+(\omega) \). The nonlinear transformation \( \exp(C^+(\omega)) - 1 \) yields \( A(\omega) \) from which \( a(z) \) is obtained by an inverse Fourier transform as \( a(z) = \mathcal{F}^{-1}\{A\}(z) \).

In the following analysis, we consider a discrete setting in which \( a_k, A_k, S_k \) and \( I_k \) denote the samples of \( a(z) \), \( A(\omega) \), \( S(\omega) \) and \( I(\omega) \) respectively.

### 3. Bias and Variance of the Cepstral Reconstruction Technique

Our model for the noisy signal is that of a deterministic signal in random additive noise \( W_k \), and is given by

\[
I_k = S_k |1 + A_k|^2 + W_k, 0 \leq k \leq N - 1, \tag{2}
\]

where \( |A_k| \ll 1 \), and where \( W_k \) is assumed to have a Gaussian distribution with zero mean and variance \( \sigma_n^2 \). Rewriting (2) as

\[
I_k = S_k |1 + A_k|^2 \left(1 + \frac{W_k}{S_k |1 + A_k|^2}\right), 0 \leq k \leq N - 1, \tag{3}
\]

and taking its logarithm, we get

\[
\log I_k = \log S_k + \log |1 + A_k|^2 + \log \left(1 + \frac{W_k}{S_k |1 + A_k|^2}\right). \tag{4}
\]

Under the assumption that the noise is small in comparison to the signal, we invoke the approximation \( \log(1 + x) = x + \mathcal{O}(x^2) \), for \( |x| < 1 \), where \( \mathcal{O} \) denotes the Landau symbol. Using the approximation, we rewrite (4) as

\[
\log I_k - \log S_k = \log |1 + A_k|^2 + \alpha_k W_k, \tag{5}
\]

where \( \alpha_k = \frac{1}{S_k |1 + A_k|^2} \), and where the higher-order terms are not shown for the sake of brevity. Taking the inverse discrete Fourier transform (IDFT) on both sides of (5), we obtain the cepstrum

\[
\hat{c}_m = \frac{1}{N} \sum_{k=0}^{N-1} \left(\log |1 + A_k|^2 + \alpha_k W_k\right) e^{j \frac{2\pi}{N} km},
\]

\[
= c_m + \frac{1}{N} \sum_{k=0}^{N} \alpha_k W_k e^{j \frac{2\pi}{N} km}, \tag{6}
\]

where \( c_m \) is the cepstrum in the absence of noise. The causal cepstrum is given by

\[
\hat{c}_m^+ = \begin{cases} c_m & \text{if } 0 \leq m \leq \frac{N}{2} - 1, \\ 0 & \text{if } \frac{N}{2} \leq m \leq N - 1. \end{cases} \tag{7}
\]

The DFT of \( \{\hat{c}_m^+, 0 \leq m \leq N - 1\} \) is

\[
\hat{C}_k^+ = \sum_{m=0}^{N-1} \hat{c}_m e^{-j \frac{2\pi}{N} km}, 0 \leq k \leq N - 1,
\]

\[
= C_k^+ + \frac{1}{N} \sum_{m=0}^{N-1} \sum_{\ell=0}^{N-1} \alpha_{\ell,m} W_{\ell} e^{j \frac{2\pi}{N} \ell m} e^{j \frac{2\pi}{N} \ell m},
\]

\[
\approx C_k^+ + \frac{1}{N} \sum_{m=0}^{N-1} \alpha_{\ell,m} W_{\ell} \sum_{\ell=0}^{N-1} e^{j \frac{2\pi}{N} \ell m} (\ell-k)^m, \tag{8}
\]

where the \( \beta_{\ell,k} \) are deterministic. When the \( \hat{C}_k^+ \) are raised to the exponent of the Napierian base \( e \), we get,

\[
\tilde{A}_k = \exp(\hat{C}_k^+) - 1,
\]

\[
= \exp(C_k^+ + \frac{1}{N} \sum_{\ell=0}^{N-1} \alpha_{\ell} \beta_{\ell,k} W_{\ell}) - 1, \tag{9}
\]

where the approximation \( e^x = 1 + x + \mathcal{O}(x^2), |x| < 1 \), has been used. This approximation is valid under the small-noise assumption. Hence,

\[
\tilde{A}_k = A_k + \exp(C_k^+) \left(\frac{1}{N} \sum_{\ell=0}^{N-1} \alpha_{\ell} \beta_{\ell,k} W_{\ell}\right). \tag{10}
\]

The scattering function estimated by applying the IDFT to the \( \{\tilde{A}_k, 0 \leq k \leq N - 1\} \) is given as

\[
\hat{a}_m = a_m + \frac{1}{N^2} \sum_{\ell=0}^{N-1} \alpha_{\ell} W_{\ell} \exp(C_k^+) \beta_{\ell,k} e^{j \frac{2\pi}{N} km}, \tag{11}
\]

\[
= a_m + \frac{1}{N^2} \sum_{\ell=0}^{N-1} \alpha_{\ell} \gamma_{\ell,m} W_{\ell}, 0 \leq m \leq \frac{N}{2} - 1,\]
where the $\gamma_{\ell,m}$ are deterministic. Applying the expectation operator $E$ on both sides of (11), we get
\[ E\{\hat{a}_m\} = a_m, \]
where we have used the assumption that the noise has zero mean. Thus, our technique yields unbiased estimates. The variance of $\hat{a}_m$ is
\[ E\{[\hat{a}_m - E\{\hat{a}_m\}]^2\} = \frac{\sigma_w^2}{N} \sum_{\ell=0}^{N-1} \alpha_{\ell}^2 |\gamma_{\ell,m}|^2. \]

4. BIAS AND VARIANCE OF THE FOURIER RECONSTRUCTION TECHNIQUE

In the conventional Fourier reconstruction technique, it is a common practice to subtract the measured source spectrum $S_k$ from the measurements $I_k$, before applying the IDFT. This operation boosts the information-carrying interference fringes. Thus, we have
\[ H_k = S_k |1 + A_k|^2 + W_k - S_k, \quad 0 \leq k \leq N - 1, \]
\[ = 2S_k \text{Re}\{A_k\} + S_k|A_k|^2 + W_k. \]
The IDFT of $\{H_k, 0 \leq k \leq N - 1\}$ is given by
\[ h_m = \frac{1}{N} \sum_{k=0}^{N-1} (2S_k \text{Re}\{A_k\} + S_k|A_k|^2 + W_k)e^{\frac{j2\pi}{N} km}. \]

Considering only the causal part of $h_m$, we get $\hat{a}_m = (s \ast a)_m + (s \ast r_{aa})_m + w_m$, where $s_m$ is the source autocorrelation, $w_m$ is the noise, and $r_{aa}$ is the autocorrelation of the scattering function, all of these being in the spatial domain. The source has a convolutional effect and the noise is additive. The expectation of $\hat{a}_m$ is given by
\[ E\{\hat{a}_m\} = (s \ast a)_m + (s \ast r_{aa})_m, \]
which implies that the estimates are biased. The variance is given by $E\{|\hat{a}_m - E\{\hat{a}_m\}|^2\} = \frac{\sigma_w^2}{N}$.

5. CRAMÉR-RAO BOUNDS

We derive the CRBs corresponding to $A_k = \sum_{m=0}^{N-1} a_m e^{-j\frac{2\pi}{N} km}$ in (2), where the $a_m$ are assumed to be real. The entries of the Fisher information matrix $J$ are given by
\[ J_{\ell,\ell'} = E\left\{ \frac{\partial \log f}{\partial a_{\ell}} \cdot \frac{\partial \log f}{\partial a_{\ell'}} \right\}, \quad 0 \leq \ell, \ell' \leq N - 1, \]
where $f$ is the probability density function of $\{I_k, 0 \leq k \leq N - 1\}$. After some computations, we arrive at
\[ J_{\ell,\ell'} = \frac{1}{4\sigma_w^2} \sum_{k=0}^{N-1} \eta_{\ell,k} \eta_{\ell',k}, \]

\[ \eta_{\ell,k} = -2S_k \cos\left(\frac{2\pi k\ell}{N}\right) - 2S_k \sum_{p=0}^{N-1} a_p \cos\left(\frac{2\pi (\ell - p)k}{N}\right). \]

\[ (J^{-1})_{k,k} \text{ gives a bound on the variance of } \hat{a}_k, \text{ however, it is not easy to compute it in a closed form. Numerical calculations show that } J \text{ becomes singular even for small values of } N. \] We propose to use the pseudoinverse of $J$ instead, keeping in mind that this may yield an inaccurate bound.

6. MONTE-CARLO ANALYSIS

Deconvolution of the source spectrum is implicit in our technique, which is not the case with the Fourier technique. Therefore, to make a fair comparison, we set the source spectrum to unity within the measurement bandwidth. We synthesize a one-dimensional scattering function corresponding to a three-layered specimen. Next, we generate 500 realizations of white Gaussian noise and compute estimates of the scattering function by using the cepstral as well as the standard Fourier techniques. We repeat the experiment for different values of signal-to-noise ratio (SNR), and corresponding to each value, we compute the variance of the estimator. Figures 2 and 3 show the variances of the estimates given by the two techniques corresponding to SNR=20 dB and 40 dB, respectively. The variances at an arbitrarily-chosen depth are shown in Fig. 4.
Fig. 3. A comparison of approximate theoretical variances vs. those obtained by simulation (SNR=40 dB). The variances of the cepstral and the Fourier techniques are nearly identical. Together with the theoretical variances and the CRBs. Note that the two techniques have almost identical variances. The pseudoinverse-based CRB is quite pessimistic, because it underestimates the performance of the best estimator. The two techniques are significantly better than the bound.

7. CONCLUSION

We addressed the noise performance of a new cepstral reconstruction algorithm for frequency-domain optical-coherence tomography [7]. We also compared the noise-sensitivity with respect to the conventional Fourier technique. Insipite of it being nonlinear, our technique has a variance that is as low as that of the linear technique. This implies a high consistency of reconstruction across different realizations of the scattering function, which is a useful feature in practice. The added advantage is that our technique suppresses the autocorrelation artifacts and yields unbiased estimates. Comparisons to iterative reconstruction techniques will be reported separately.

8. REFERENCES


