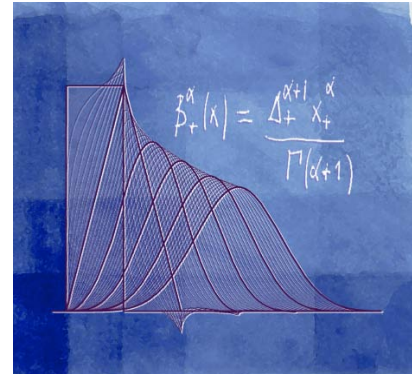


Sparse stochastic processes: A statistical framework for modern signal processing

Michael Unser
Biomedical Imaging Group
EPFL, Lausanne, Switzerland



Plenary talk, Int. Conf. Syst. Sig. Im. Proc. (IWSSIP), Bucharest, July 7-9, 2013

20th century statistical signal processing

Hypothesis: Signal = stationary **Gaussian** process

Karhunen-Loève transform (KLT) is optimal for compression

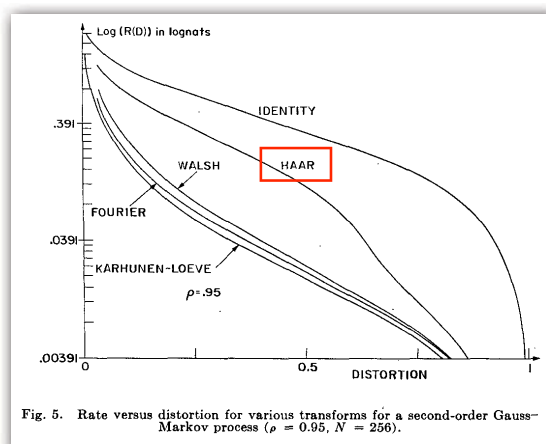


Fig. 5. Rate versus distortion for various transforms for a second-order Gauss-Markov process ($\rho = 0.95$, $N = 256$).

(Pearl et al., *IEEE Trans. Com* 1972)

DCT asymptotically equivalent to KLT

(Ahmed-Rao, 1975; U., 1984)



20th century statistical signal processing (cont'd)

Hypothesis: Signal = **Gaussian** process

$$\mathbf{y} = \mathbf{H}\mathbf{s} + \mathbf{n}$$

Noise: i.i.d. Gaussian with variance σ^2

Signal covariance: $\mathbf{C}_s = \mathbb{E}\{\mathbf{s} \cdot \mathbf{s}^T\}$

Wiener filter is **optimal** for restoration/denoising

$$\mathbf{s}_{\text{LMMSE}} = \mathbf{C}_s \mathbf{H}^T (\mathbf{H} \mathbf{C}_s \mathbf{H}^T + \sigma^2 \mathbf{I})^{-1} \mathbf{y} = \mathbf{F}_{\text{Wiener}} \mathbf{y}$$

Wiener (LMMSE) solution = Gauss MMSE = Gauss MAP

$$\mathbf{s}_{\text{MAP}} = \arg \min_{\mathbf{s}} \underbrace{\frac{1}{\sigma^2} \|\mathbf{y} - \mathbf{H}\mathbf{s}\|_2^2}_{\text{Data Log likelihood}} + \underbrace{\|\mathbf{C}_s^{-1/2} \mathbf{s}\|_2^2}_{\text{Gaussian prior likelihood}}$$

\Leftrightarrow quadratic regularization (Tikhonov)

3

Then came wavelets ... and sparsity



Stéphane Mallat



Ingrid Daubechies

Alfred Haar

1910



Martin Vetterli

1982

1987-88

1994



David Donoho

Sparsity

2006



Emmanuel Candès

Applications

Compressed sensing

4

Fact 1: Wavelets can outperform Wiener filter

MAGNETIC RESONANCE IN MEDICINE 21, 288-295 (1991)

COMMUNICATIONS

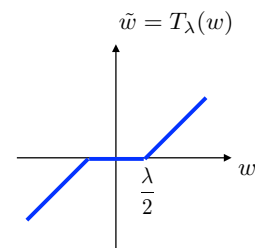
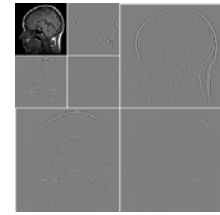
Filtering Noise from Images with Wavelet Transforms

J. B. WEAVER,* YANSUN XU,* D. M. HEALY, JR.,† AND L. D. CROMWELL*

* Department of Radiology, Dartmouth-Hitchcock Medical Center; and † Department of Mathematics, Dartmouth College, Hanover, New Hampshire 03755

Received April 12, 1991

A new method of filtering MR images is presented that uses wavelet transforms instead of Fourier transforms. The new filtering method does not reduce the sharpness of edges. However, the new method does eliminate any small structures that are similar in size to the noise eliminated. There are many possible extensions of the filter. © 1991 Academic Press, Inc.



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Fact 2: Wavelet coding can outperform jpeg

$$f(\mathbf{x}) = \sum_{i,k} \psi_{i,k}(\mathbf{x}) w_{i,k}$$



66.4 dB

Wavelet transform



0.00%

Inverse wavelet transform

Discarding "small coefficients"

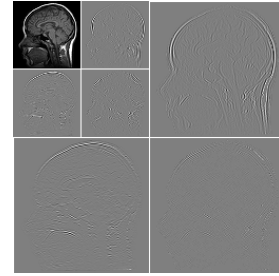
(Shapiro, IEEE-IP 1993)



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Fact 3: l_1 schemes can outperform l_2 optimization

$$s^* = \underset{s}{\operatorname{argmin}} \underbrace{\|y - \mathbf{H}s\|_2^2}_{\text{data consistency}} + \underbrace{\lambda \mathcal{R}(s)}_{\text{regularization}}$$



■ Wavelet-domain regularization

Wavelet expansion: $s = \mathbf{W}v$ (typically, sparse)

Wavelet-domain sparsity-constraint: $\mathcal{R}(s) = \|v\|_{\ell_1}$ with $v = \mathbf{W}^{-1}s$

Iterated shrinkage-thresholding algorithm (ISTA, FISTA)

(Nowak et al., Daubechies et al. 2004)

■ ℓ_1 regularization (Total variation)

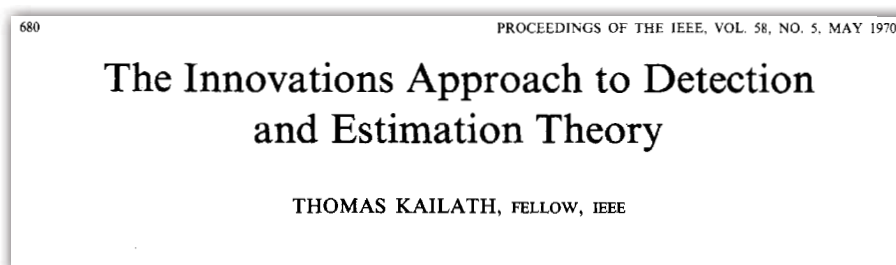
$\mathcal{R}(s) = \|\mathbf{L}s\|_{\ell_1}$ with \mathbf{L} : gradient

(Rudin-Osher, 1992)

Iterative reweighted least squares (IRLS) or FISTA

7

Quest for a unifying framework: the precursor



■ Gaussian stationary processes as a filtered white noise

$$s(t) = (h * w)(t) \quad \Phi_s(\omega) = |H(\omega)|^2 \Phi_w(\omega) \propto |H(\omega)|^2$$

Frequency response (shaping filter): $H(\omega) = \int_{\mathbb{R}} h(t) e^{-j\omega t} dt$

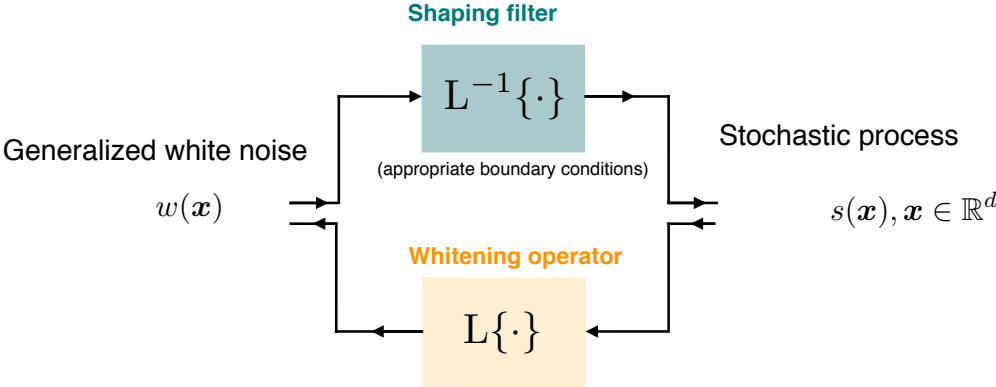
■ Whitening operator

$$\text{Innovation: } Ls(t) = w(t) \quad L(\omega) = \frac{1}{H(\omega)}$$

w is **Gaussian** stationary and independent at every point

8

Continuous-domain innovation model



Our work: the **non-gaussian** part of this story

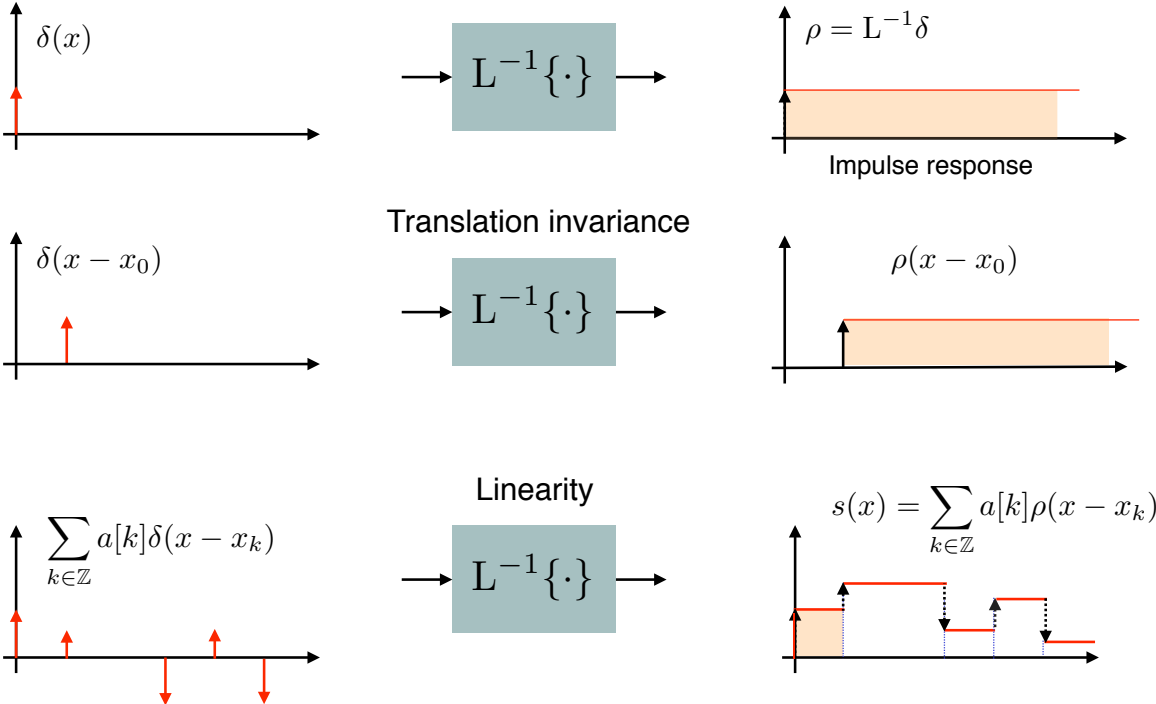
Main finding: it is **necessarily** sparse (*infinitely divisible*)

Why? ... as explained by our current research ...

(invoking powerful theorems in functional analysis: Bochner-Minlos, Gelfand, Schoenberg & Lévy-Khinchine)

Innovation-based synthesis of splines


$$L = \frac{d}{dx} = D \Rightarrow L^{-1}: \text{integrator}$$



OUTLINE

- Sparse stochastic processes
 - Generalized innovation model
 - Gelfand's theory of generalized stochastic processes
 - Statistical characterization of sparse stochastic processes
 - Lévy processes and their generalization
 - Fractal processes: Gaussian vs. sparse
- Applications
 - Modeling of signals (audio)
 - Algorithms for sparse signal recovery as MAP estimators
 - Optimal denoising (MMSE)
 - Sparse representations, optimal transforms

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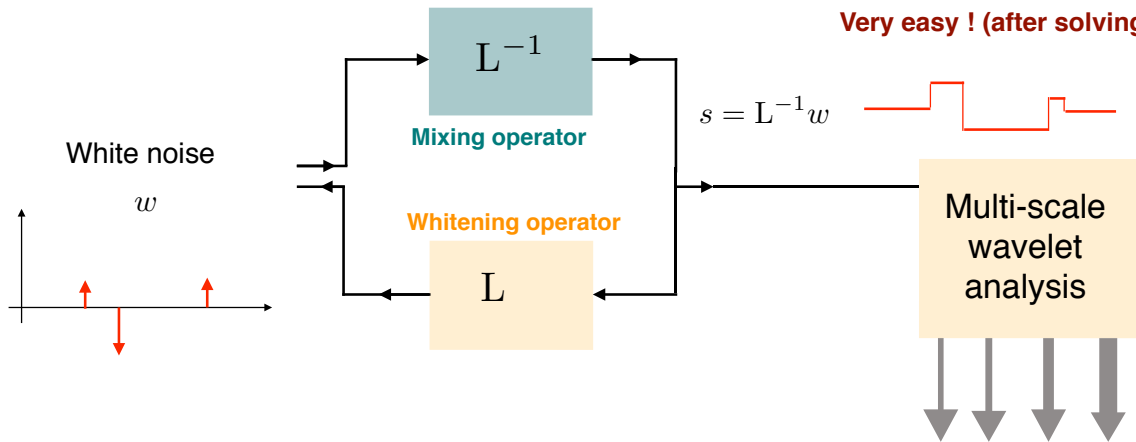
Sparse stochastic
processes

12

Road map for theory of sparse processes

② Specification of inverse operator
Functional analysis solution of SDE

③ Characterization of generalized stochastic process
Very easy ! (after solving 1. & 2.)



① Characterization of continuous-domain white noise

Higher mathematics: **generalized functions (Schwartz)**
measures on topological vector spaces

Gelfand's theory of *generalized stochastic processes*
Infinite divisibility (Lévy-Khintchine formula)

④ Characterization of transform-domain statistics

Easy when: $\psi_i = L^* \phi_i$

Generalized innovation process

■ Difficulty 1: $w \neq w(x)$ is too rough to have a pointwise interpretation



■ Difficulty 2: w is an infinite-dimensional random entity;
 its "pdf" can be formally specified by a measure $\mathcal{P}_w(E)$ where $E \subseteq S'(\mathbb{R}^d)$

■ Axiomatic definition (Gelfand-Vilenkin 1964)

w is a generalized innovation process (or continuous-domain white noise) over $S'(\mathbb{R}^d)$ if

1. **Observability** : $X = \langle w, \varphi \rangle$ is a well-defined random variable for any test function $\varphi \in \mathcal{S}(\mathbb{R}^d)$.
2. **Stationarity** : $X_{x_0} = \langle w, \varphi(\cdot - x_0) \rangle$ is identically distributed for all $x_0 \in \mathbb{R}^d$.
3. **Independent atoms** : $X_1 = \langle w, \varphi_1 \rangle$ and $X_2 = \langle w, \varphi_2 \rangle$ are independent whenever φ_1 and φ_2 have non-intersecting support.

$$X_1 = \langle \text{[blue waveform]}, \text{[red waveform]} \rangle$$

$$X_2 = \langle \text{[blue waveform]}, \text{[red waveform]} \rangle$$

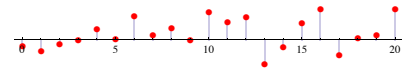
■ Characteristic functional ($\omega \rightarrow \varphi$)

$$\widehat{\mathcal{P}}_w(\varphi) = \mathbb{E}\{e^{j\langle w, \varphi \rangle}\} = \int_{S'} e^{j\langle s, \varphi \rangle} \mathcal{P}_w(ds)$$

Defining Gaussian noise: discrete vs. continuous

Lévy exponent: $\log \hat{p}_X(\omega) = f(\omega) = -\frac{1}{2}\omega^2$

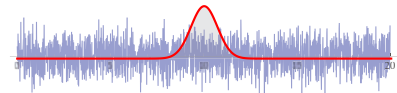
■ Discrete white Gaussian noise



$X = (X_1, \dots, X_N)$ with X_n i.i.d standardized Gaussian

Characteristic function: $\hat{p}_X(\omega) = g(\omega) = \exp\left(\sum_{n=1}^N f(\omega_n)\right) = e^{-\frac{1}{2}\|\omega\|^2}$

■ Continuous-domain white Gaussian noise



Infinite-dimensional entity w with generic observations $X_n = \langle w, \varphi_n \rangle$

Characteristic functional: $\widehat{\mathcal{P}}_w(\varphi) = G(\varphi) = e^{-\frac{1}{2}\|\varphi\|_{L_2}^2} = \exp\left(\int_{\mathbb{R}} f(\varphi(x)) dx\right)$

$\hat{p}_{X_n}(\omega) = \mathbb{E}\{e^{j\omega\langle w, \varphi_n \rangle}\} = \mathbb{E}\{e^{j\langle w, \omega\varphi_n \rangle}\} = \widehat{\mathcal{P}}_w(\omega\varphi_n) = e^{-\frac{1}{2}\omega^2\|\varphi_n\|_{L_2}^2}$

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Characterization of generalized innovation

$$\begin{aligned} X_\varphi = \langle w, \varphi \rangle &= \langle \text{[noise]}, \text{[smooth curve]} \rangle \triangleq \lim_{n \rightarrow \infty} \langle \text{[noise]}, \text{[staircase]} \rangle \\ &= \lim_{n \rightarrow \infty} \langle \text{[noise]}, \text{[rect]} \rangle + \dots + \langle \text{[noise]}, \text{[rect]} \rangle \end{aligned}$$

Theorem

Let w be a generalized stochastic process such that $X_{\text{id}} = \langle w, \text{rect} \rangle$ is well-defined. Then, w is a generalized innovation (white noise) over $S'(\mathbb{R}^d)$ **if and only if** its characteristic form is given by

$$\widehat{\mathcal{P}}_w(\varphi) = \mathbb{E}\{e^{j\langle w, \varphi \rangle}\} = \exp\left(\int_{\mathbb{R}^d} f(\varphi(x)) dx\right)$$

where $f(\omega)$ is a **valid Lévy exponent** (in fact, the Lévy exponent of X_{id}).

Moreover, the random variables $X_\varphi = \langle w, \varphi \rangle$ are all **infinitely divisible** with modified Lévy exponent

$$f_\varphi(\omega) = \int_{\mathbb{R}^d} f(\omega\varphi(x)) dx$$

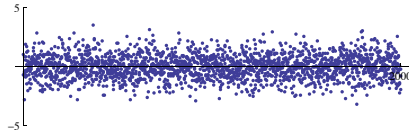
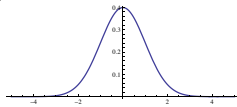
(Gelfand-Vilenkin 1964; Amini-U. under review)

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Examples of iid noise distributions

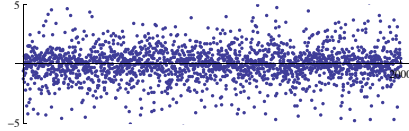
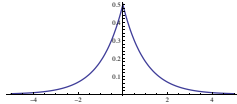
$p_{\text{id}}(x)$ Observations: $X_n = \langle w, \text{rect}(\cdot - n) \rangle$

(a) Gaussian



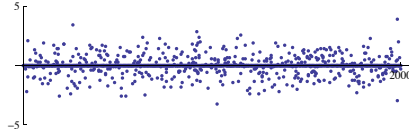
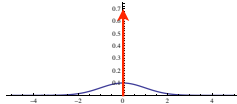
$$f(\omega) = -\frac{1}{2\sigma_0^2}|\omega|^2$$

(b) Laplace



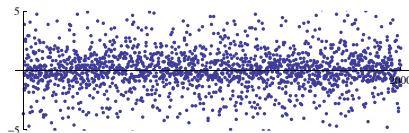
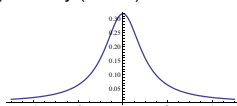
$$f(\omega) = \log\left(\frac{1}{1+\omega^2}\right)$$

(c) Compound Poisson



$$f(\omega) = \lambda \int_{\mathbb{R}} (e^{jx\omega} - 1) p(x) dx$$

(d) Cauchy (stable)

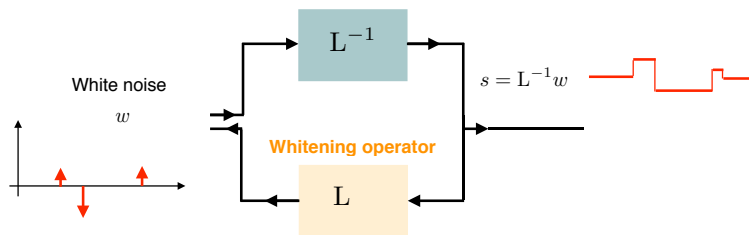


$$f(\omega) = -s|\omega|$$

Sparser

Complete mathematical characterization: $\widehat{\mathcal{P}}_w(\varphi) = \exp\left(\int_{\mathbb{R}^d} f(\varphi(\mathbf{x})) d\mathbf{x}\right)$

Steps 2 + 3: Characterization of sparse process



■ Abstract formulation of innovation model

$$s = L^{-1}w \Leftrightarrow \forall \varphi \in \mathcal{S}, \quad \langle \varphi, s \rangle = \langle \varphi, L^{-1}w \rangle = \langle \underbrace{L^{-1*}\varphi}_s, w \rangle$$

$$\Rightarrow \widehat{\mathcal{P}}_s(\varphi) = \mathbb{E}\{e^{j\langle s, \varphi \rangle}\} = \widehat{\mathcal{P}}_w(L^{-1*}\varphi) = \exp\left(\int_{\mathbb{R}^d} f(L^{-1*}\varphi(\mathbf{x})) d\mathbf{x}\right)$$

Sufficient condition for existence:

$$L^{-1*} \text{ continuous operator: } \mathcal{S}(\mathbb{R}^d) \rightarrow L_p(\mathbb{R}^d)$$

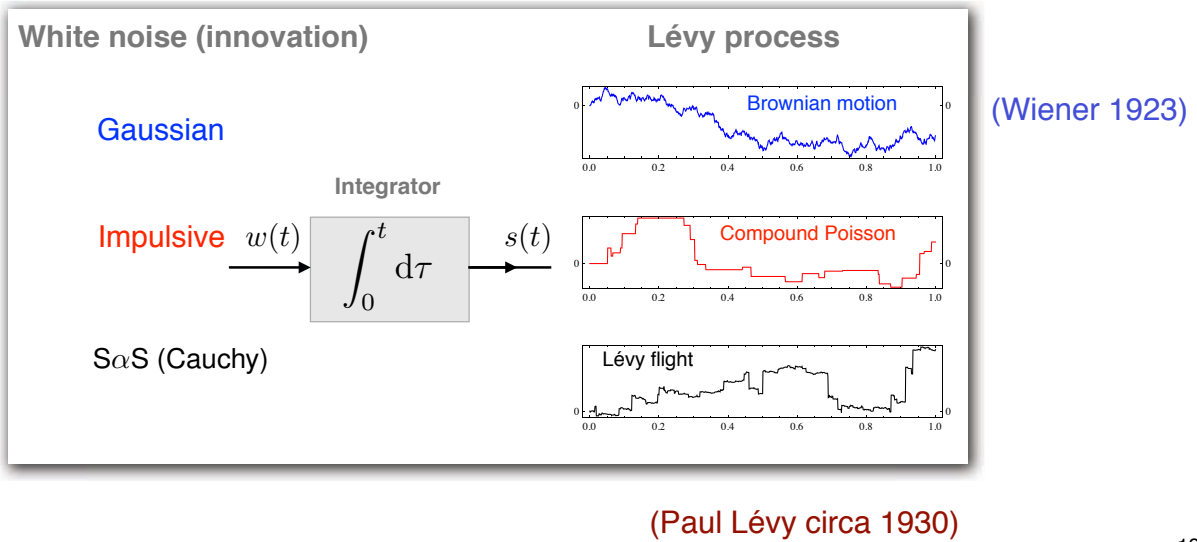
(U.-Tafti-Sun, preprint ArXiv 2011)

Example 1: Lévy processes

$$Ds = w \quad (\text{unstable SDE !})$$

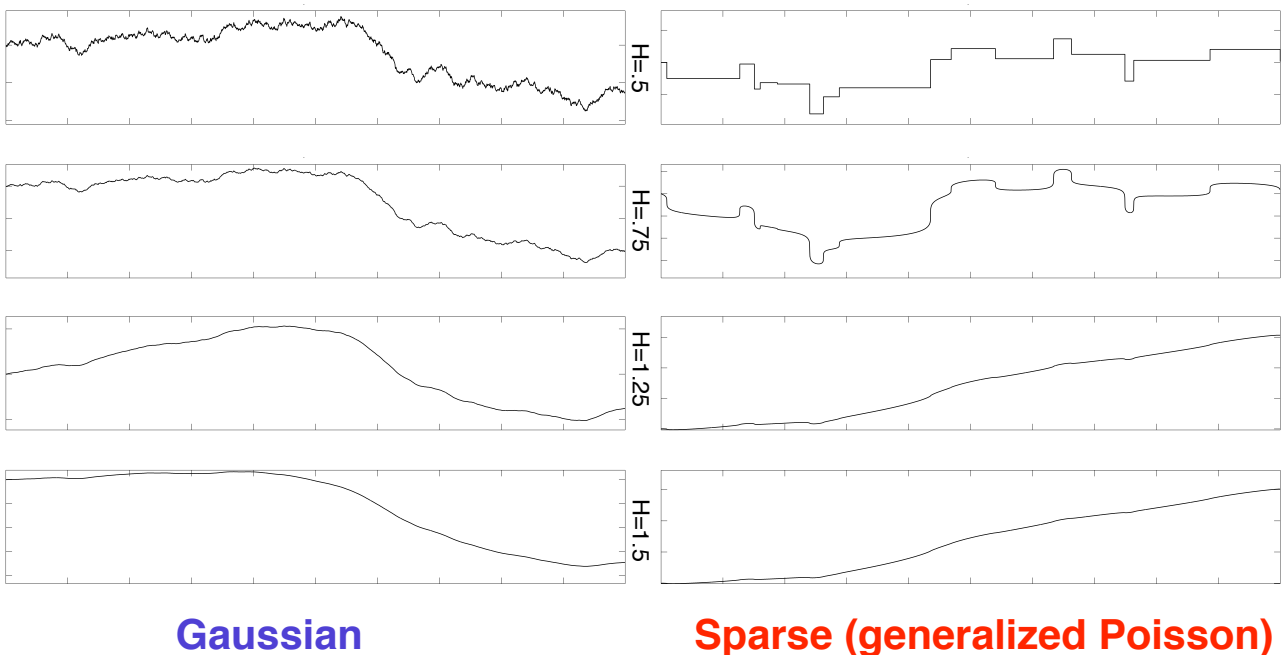
$$s = D_0^{-1}w \Leftrightarrow \forall \varphi \in \mathcal{S}, \quad \langle \varphi, s \rangle = \langle D_0^{-1*}\varphi, w \rangle$$

$$D_0^{-1}\varphi(t) = \int_0^t \varphi(\tau) d\tau$$



Example 2: Self-similar processes

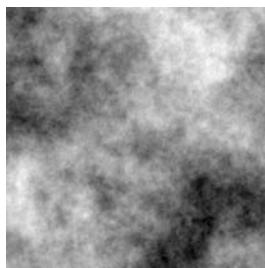
$$L \xleftrightarrow{\mathcal{F}} (j\omega)^{H+\frac{1}{2}} \Rightarrow L^{-1}: \text{fractional integrator}$$



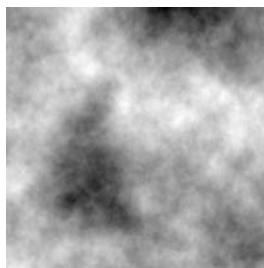
Scale- and rotation-invariant processes

Stochastic partial differential equation : $(-\Delta)^{\frac{H+1}{2}} s(\mathbf{x}) = w(\mathbf{x})$

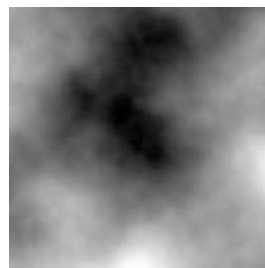
Gaussian



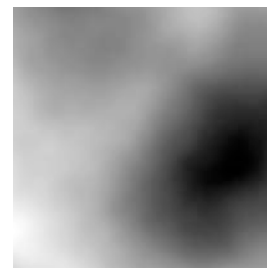
H=0.5



H=0.75

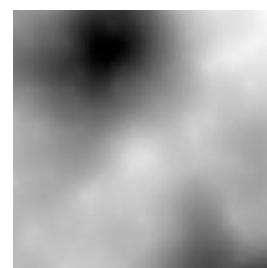
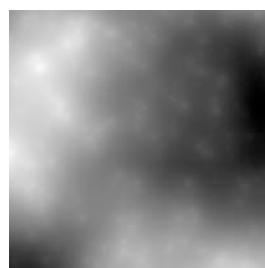
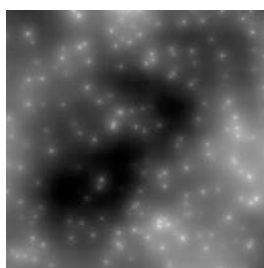
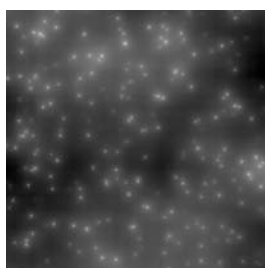


H=1.25



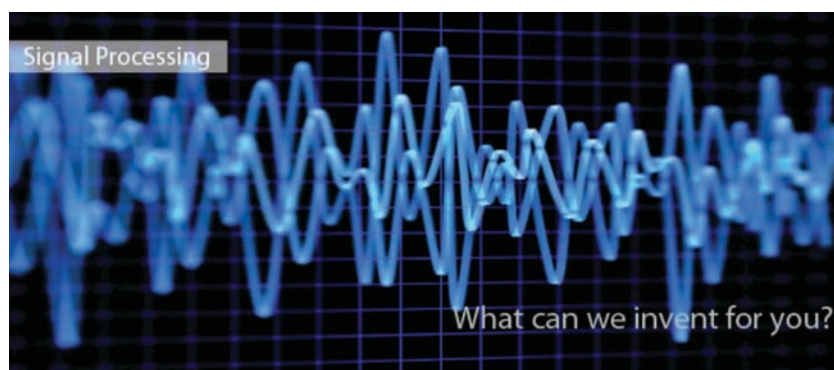
H=1.75

Sparse (generalized Poisson)



(U.-Tafti, *IEEE-SP* 2010)

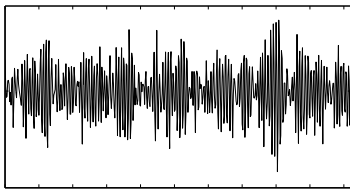
A brief panorama of applications



A1. Signal modeling (Audio)

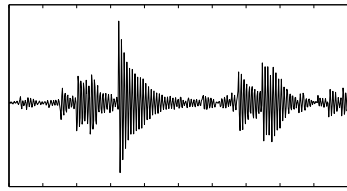
- Sparse, bandpass processes

poles = [-.05 + jπ/2, -.05 - jπ/2], zeros = []



(a) Gaussian

$$L = \frac{d^n}{dt^n} + a_{n-1} \frac{d^{n-1}}{dt^{n-1}} + \dots + a_1 \frac{d}{dt} + a_0 I$$



(b) Alpha stable $\alpha=1.2$

- Mixed sparse processes: $s_{\text{mix}} = s_1 + \dots + s_M$

$$\widehat{\mathcal{P}}_{s_{\text{mix}}}(\varphi) = \prod_{m=1}^M \widehat{\mathcal{P}}_{s_m}(\varphi) = \exp\left(\int_{\mathbb{R}} \sum_{m=1}^M f_m(L_m^{-1*} \varphi(t)) dt\right)$$



Gaussian (Am)



generalized Lévy (Am, SαS)

A2. Biomedical imaging: MAP reconstruction

- Innovation model of the signal

$$\begin{aligned} Ls &= w \\ s &= L^{-1}w \end{aligned}$$

- Signal decoupling: discrete version of operator

$$u(x) = L_d s(x) \quad \Leftrightarrow \quad \mathbf{u} = \mathbf{L} \mathbf{s} \quad (\text{matrix notation})$$

↑

Generalized B-spline
 $\beta_L = L_d L^{-1} \delta$

←

- Statistical characterization

- $X = [\mathbf{u}]_n$ identically distributed (approx. independent)
- Probability density function: $p_X(x) = \mathcal{F}^{-1}\{\widehat{\mathcal{P}}_w(\omega \beta_L)\}(x)$
- Potential function: $\Phi_X(x) = -\log p_X(x)$

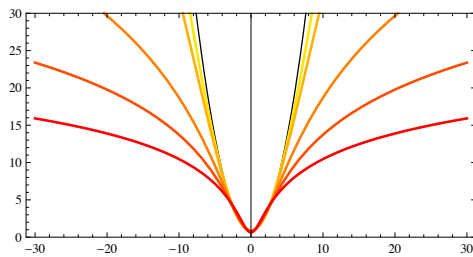
- Maximum a posteriori (MAP) estimator for AWN

$$s^* = \operatorname{argmin} \left(\frac{1}{2} \|\mathbf{g} - \mathbf{H}\mathbf{s}\|_2^2 + \sigma^2 \sum_n \Phi_X([\mathbf{L}\mathbf{s}]_n) \right)$$

MAP estimator: special cases

$$\mathbf{s}^* = \operatorname{argmin} \left(\frac{1}{2} \|\mathbf{g} - \mathbf{H}\mathbf{s}\|_2^2 + \sigma^2 \sum_n \Phi_X([\mathbf{L}\mathbf{s}]_n) \right)$$

- Sparsier
- Gaussian: $p_X(x) = \frac{1}{\sqrt{2\pi}\sigma_0} e^{-x^2/(2\sigma_0^2)} \Rightarrow \Phi_X(x) = \frac{1}{2\sigma_0^2} x^2$
 - Laplace: $p_X(x) = \frac{\lambda}{2} e^{-\lambda|x|} \Rightarrow \Phi_X(x) = \lambda|x|$
 - Student: $p_X(x) = \frac{1}{B(r, \frac{1}{2})} \left(\frac{1}{x^2 + 1} \right)^{r+\frac{1}{2}} \Rightarrow \Phi_X(x) = \left(r + \frac{1}{2} \right) \log(1 + x^2)$



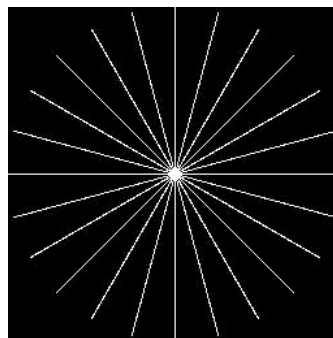
Student potentials: $r = 2, 4, 8, 32$ (fixed variance)

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MRI: Shepp-Logan phantom



Original SL Phantom



Fourier Sampling Pattern
12 Angles



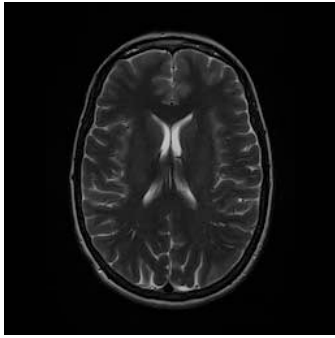
Laplace prior (TV)



Student prior (log)

L : gradient
Optimized parameters

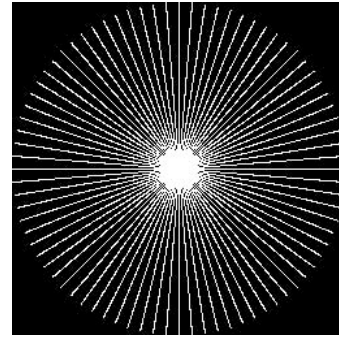
MRI reconstruction



Real T2 Brain Image



MR Angiography Image



k-space sampling pattern

40 radial lines

L : gradient

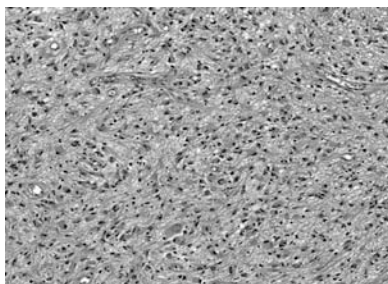
Optimized parameters

Reconstruction results in dB

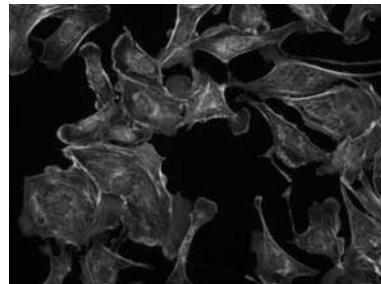
	Gaussian Estimator	Laplace Estimator	Student's Estimator
T2 brain Image	8.71	16.08	15.79
MR Angiography Image	6.31	14.48	14.97

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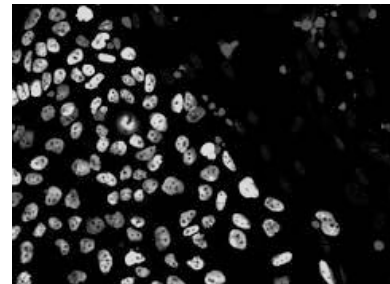
2D deconvolution experiment



Astrocytes cells



bovine pulmonary artery cells



human embryonic stem cells

Disk shaped PSF (7x7)

L : gradient

Deconvolution results in dB

Optimized parameters

	Gaussian Estimator	Laplace Estimator	Student's Estimator
Astrocytes cells	12.18	10.48	10.52
Pulmonary cells	16.90	19.04	18.34
Stem cells	15.81	20.19	20.50

(Bostan et al., *IEEE-IP* 2013)

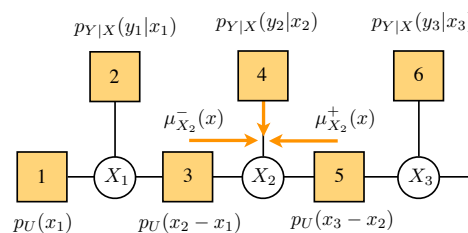
A3. Optimal denoising: MMSE formulation

- Measurement model: $\mathbf{y} = \mathbf{x} + \mathbf{n}$

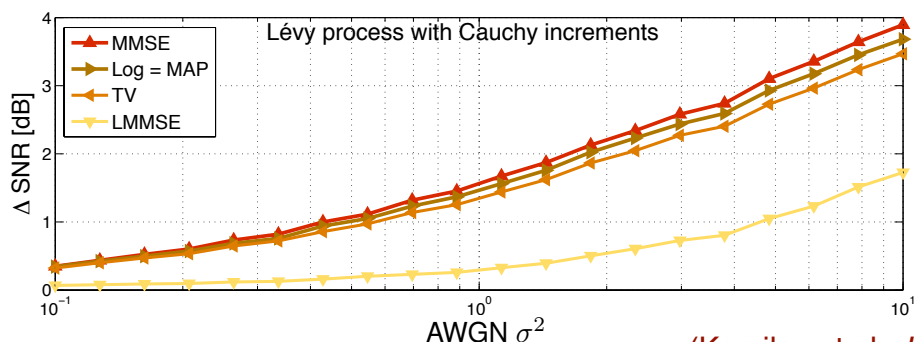
\mathbf{n} : additive white Gaussian noise
 \mathbf{x} : sparse first-order process (e.g., Lévy flight)
 $\mathbf{u} = \mathbf{L}\mathbf{s}$: discrete innovation (i.i.d. and infinite divisible)

Optimal estimator: $\mathbf{x}_{\text{MMSE}} = E\{\mathbf{x}|\mathbf{y}\}$

$$p(\mathbf{x}|\mathbf{y}) = \frac{1}{Z} \prod_{n=1}^N p_{Y|X}(y_n|x_n) \prod_{n=1}^N p_U(x_n - x_{n-1})$$



Belief propagation algorithm



(Kamilov et al., *IEEE-SP* 2013)

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A4. Sparse representations, optimal transforms

- Innovation model (SDE)

$$\begin{aligned}
 \mathbf{L}\mathbf{s} &= \mathbf{w} \\
 \mathbf{s} &= \mathbf{L}^{-1}\mathbf{w}
 \end{aligned}$$

- Admissible basis function: $\psi_{i,k} = \mathbf{L}^* \phi_{i,k}$ with $\phi_{i,k} \in L_p(\mathbb{R}^d)$

- Signal expansion = equivalent white-noise analysis

$$Y = \langle \psi_{i,k}, \mathbf{s} \rangle = \langle \mathbf{L}^* \phi_{i,k}, \mathbf{L}^{-1} \mathbf{w} \rangle = \langle \phi_{i,k}, \mathbf{w} \rangle$$

$$\implies \hat{p}_Y(\omega) = \widehat{\mathcal{P}}_w(\omega \phi_{i,k})$$

- Statistical implications

- Transform-domain pdfs are infinitely divisible
- Quality of decoupling depends upon support of wavelet/smoothing kernel $\phi_{i,k}$

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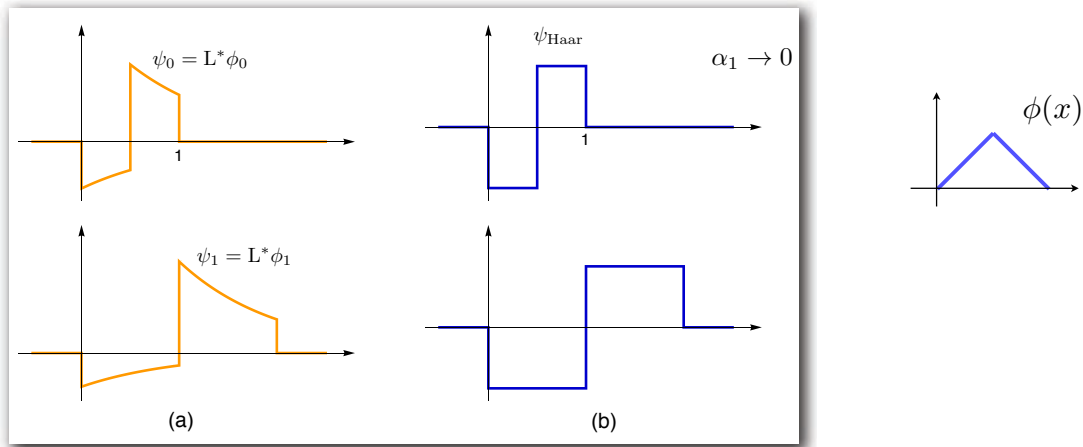
Operator-like wavelets for sparse AR(1) processes

Innovation model: $Ls = w \Leftrightarrow s = L^{-1}w$ with $L = (D - \alpha_1 I)$

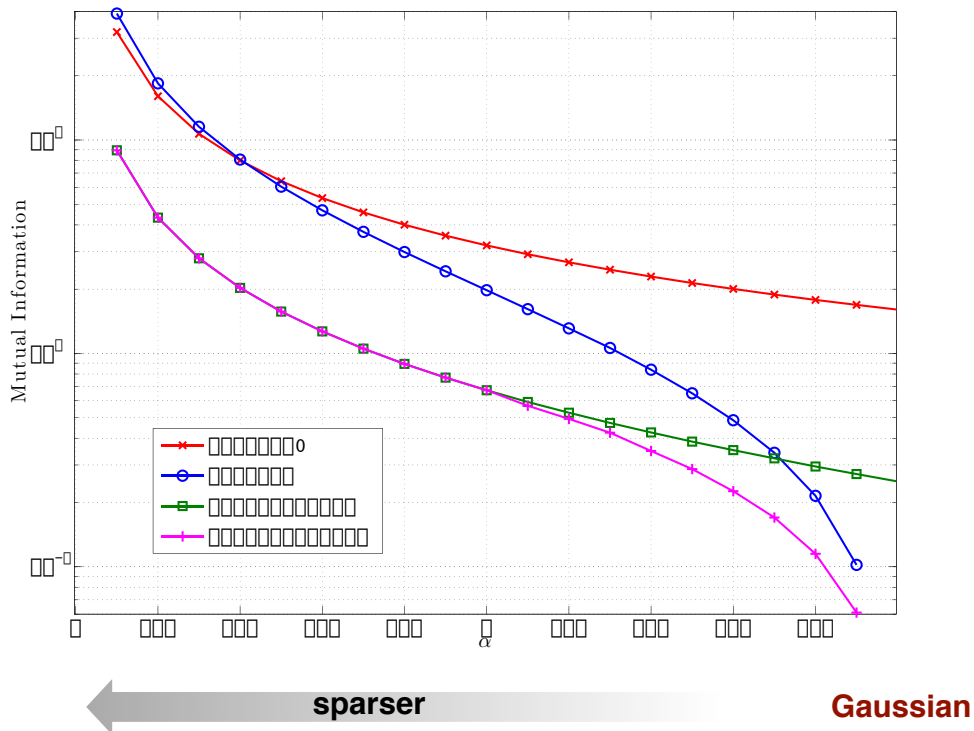
Operator-like wavelet: $\psi_i = L^* \phi_i$ with ϕ_i : smoothing kernel

(Khalidov-U., 2006)

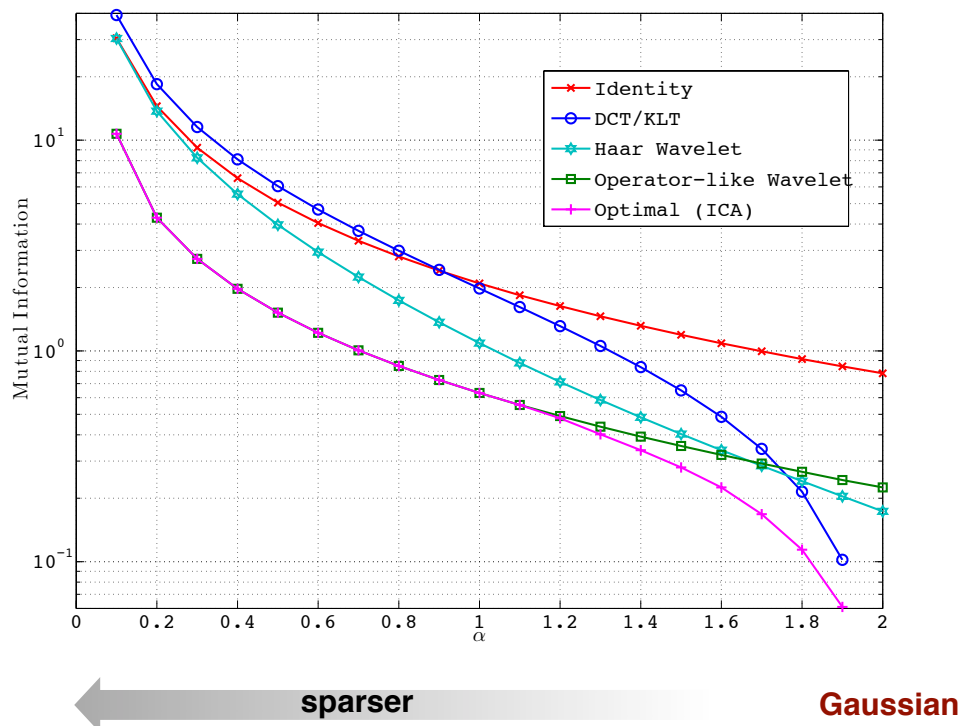
Wavelet analysis: $\langle s, \psi_i(\cdot - t_0) \rangle = \langle L^{-1}w, L^* \phi_i(\cdot - t_0) \rangle = \langle w, \phi_i(\cdot - t_0) \rangle$



Orthogonal expansion of a $S_{\alpha}S$ Lévy process



Orthogonal expansion of a $S_{\alpha}S$ AR(1) process



(Pad-U. ICASSP'13, SPARS'13)

$e^{\alpha_1} = 0.9, M = 64$

CONCLUSION

- Unifying continuous-domain innovation model
 - Backward compatibility with classical Gaussian theory
 - Operator-based formulation: Lévy-driven SDEs
 - **Gaussian** vs. **sparse** (generalized Poisson, student, $S_{\alpha}S$)
 - Focus on unstable SDEs \Rightarrow non-stationary, self-similar processes
- Wavelet analysis vs. regularization
 - Central role of B-spline (see papers)
 - Sparsification/decoupling via “operator-like” behavior
- Theoretical framework for sparse signal recovery
 - Analytical determination of PDF in **any** transformed domain
 - Predictive power: transform coding/denoising (**facts 1, 2, 3**)
 - New statistically-founded sparsity priors
 - Derivation of estimators (MAP vs. MMSE): link with LASSO and l_1 methods for sparse signal recovery (Compressed sensing)

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■ Recent applications

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Acknowledgments

Many thanks to

- **Dr. Pouya Tafti**
- Prof. Qiyu Sun
- Prof. Thierry Blu
- Emrah Bostan
- Dr. Matthieu Guerquin-Kern
- Dr. Arash Amini
- Dr. Hagai Kirshner
- Pedram Pad
- Ulugbek Kamilov
- Masih Nilchian



- **Members of EPFL's Biomedical Imaging Group**



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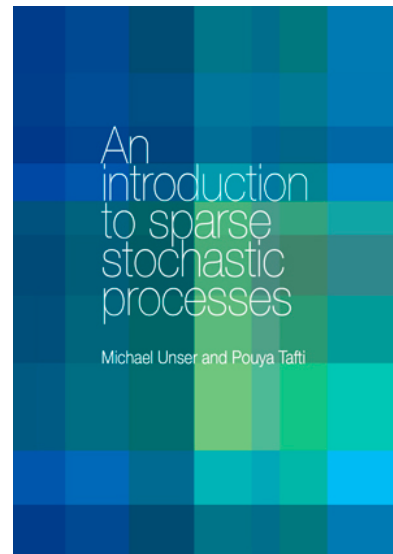
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- ▶ Sparse signal recovery
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<http://www.sparseprocesses.org>